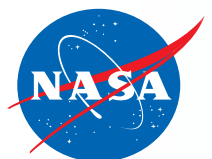


# **Characterization of Acoustic Liners at High Frequencies in a Laboratory Environment**

**Alexander N. Carr**  
**NASA Langley Research Center, Hampton, VA**

**Acoustics Liner Workshop**  
**Hampton, VA**  
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# Acknowledgments

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Portions of this presentation focus on experiments led by Mike Jones and Chelsea Solano with contributions from Larry Becker and Doug Nark

# Motivation

- 2020 survey indicates liner characterization up to 5 – 6 kHz may be desired
- Liners traditionally tested  $\leq 3$  kHz in LTF [1]
  - Plenty of experimental studies in literature at plane-wave frequencies [2–9]
  - Impedance models validated in the plane-wave regime
  - Lack of background information regarding high frequency impedance testing [10, 11]
- Difficulties associated with high frequency testing that we expected
  - Achieving sufficient SPL at high frequency
  - Different analysis techniques, higher order modes (HOMs) present
- This talk will review
  - Our current progress in high frequency testing in LTF
  - How we have approached expected difficulties
  - How we plan to approach unexpected difficulties

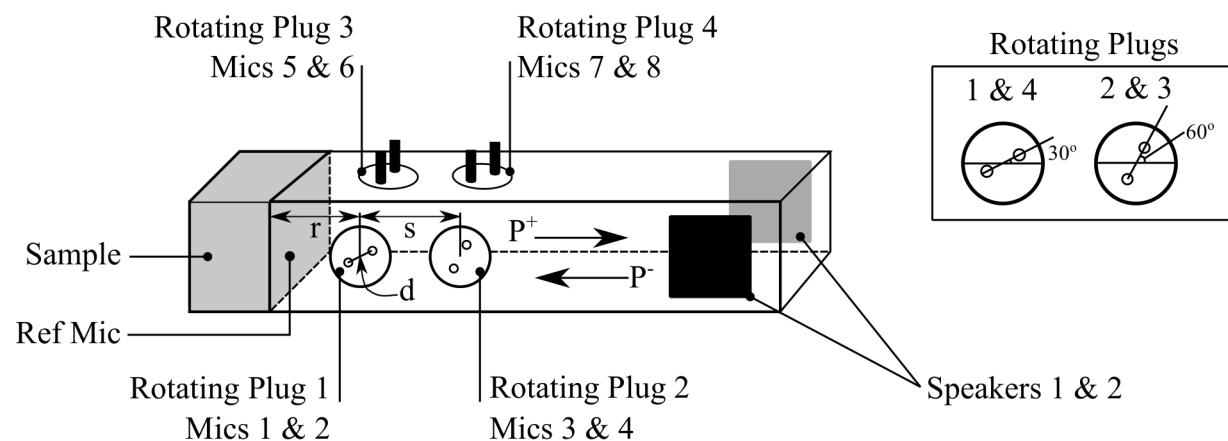
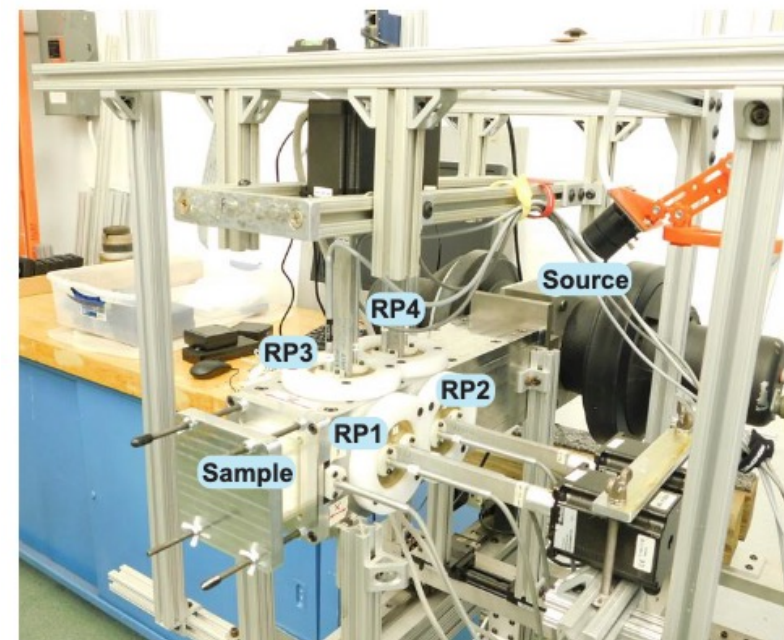




# Background: High Intensity Modal Impedance Tube

## High Intensity Modal Impedance Tube (HIMIT)

- No flow
- 2'' $\times$ 2'' rectangular waveguide
- Source generation
  - Two compression drivers (155+ dB)
  - Hartmann generator (170 dB)
- 14'' in length
- 0.25'' reference mic, 0.25'' standoff distance from sample face
- 4 rotating plugs, each with 2 mics (0.125'')
- Two-microphone method (plane wave)
- Modal impedance eduction (higher frequencies)
  - Requires microphone calibration
  - Performed in situ with CSQ3 sample





# Impedance Eduction with Modal Analysis

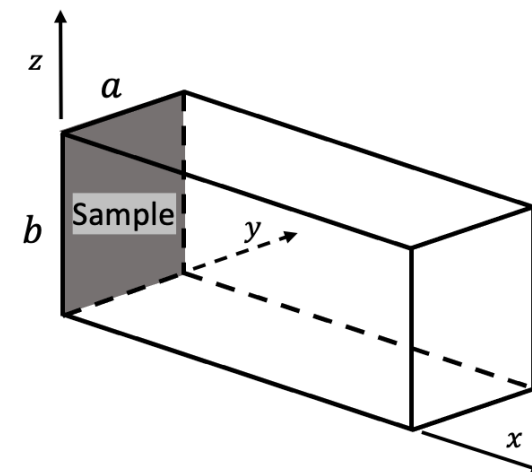
## Modal Analysis

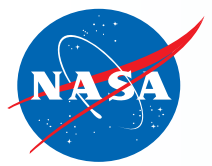
$$P(x, y, z) = \sum_{n=0}^{\infty} \sum_{m=0}^{\infty} [A_{nm}^{+} e^{-i\gamma_{nm}x} + A_{nm}^{-} e^{i\gamma_{nm}x}] \Psi_{nm}(y, z)$$

- $\Psi_{nm}(y, z) = \cos\left(\frac{n\pi y}{a}\right) \cos\left(\frac{m\pi z}{b}\right)$
- 8 mics, take the absolute SPL and phase at each mic
- Solve for the modal amplitudes using the relation:  $\mathbf{A} = [\mathbf{T}^T \mathbf{T}]^{-1} \mathbf{T}^T \mathbf{P}$

$$\mathbf{P} = \begin{bmatrix} P(x_1, y_1, z_1) \\ \vdots \\ P(x_8, y_8, z_8) \end{bmatrix} \quad \mathbf{T} = \begin{bmatrix} T_{00}^{+}(x_1, y_1, z_1) & \cdots & T_{NM}^{+}(x_1, y_1, z_1) & T_{00}^{-}(x_1, y_1, z_1) & \cdots & T_{NM}^{-}(x_1, y_1, z_1) \\ \vdots & \ddots & \vdots & \vdots & \ddots & \vdots \\ T_{00}^{+}(x_8, y_8, z_8) & \cdots & T_{NM}^{+}(x_8, y_8, z_8) & T_{00}^{-}(x_8, y_8, z_8) & \cdots & T_{NM}^{-}(x_8, y_8, z_8) \end{bmatrix} \quad \mathbf{A} = \begin{bmatrix} A_{00}^{+} \\ \vdots \\ A_{NM}^{+} \\ A_{00}^{-} \\ \vdots \\ A_{NM}^{-} \end{bmatrix}$$

- $T_{nm}^{+}(x_i, y_i, z_i) = e^{-i\gamma_{nm}x_i} \Psi_{nm}(y_i, z_i)$  and  $T_{nm}^{-}(x_i, y_i, z_i) = e^{i\gamma_{nm}x_i} \Psi_{nm}(y_i, z_i)$
- $[\mathbf{T}^T \mathbf{T}]^{-1} \mathbf{T}^T$  is computed using singular value decomposition
- Method is **most accurate** for predicting the **dominant mode**





# Impedance Eduction with Modal Analysis

## Impedance Eduction

- Normalized specific acoustic impedance of the sample

$$\zeta = -\frac{P}{\rho c U_x} = -k \frac{\sum_{n=0}^{\infty} \sum_{m=0}^{\infty} [A_{nm}^+ e^{-ik_x x} + A_{nm}^- e^{ik_x x}] \Psi_{nm}(y, z)}{\sum_{n=0}^{\infty} \sum_{m=0}^{\infty} \gamma_{nm} [A_{nm}^+ e^{-ik_x x} - A_{nm}^- e^{ik_x x}] \Psi_{nm}(y, z)}$$

- Impedance for a single mode

$$\zeta_{nm,nm} = \frac{k}{\gamma_{nm}} \frac{1 + R_{nm,nm}}{1 - R_{nm,nm}}$$

(1)

- $R_{nm,nm} = A_{nm}^+ / A_{nm}^-$
- Accurate for the dominant mode with large separation from other modes
- Impedance should be computed for mode with highest power

# Investigations with HIMIT (so far)

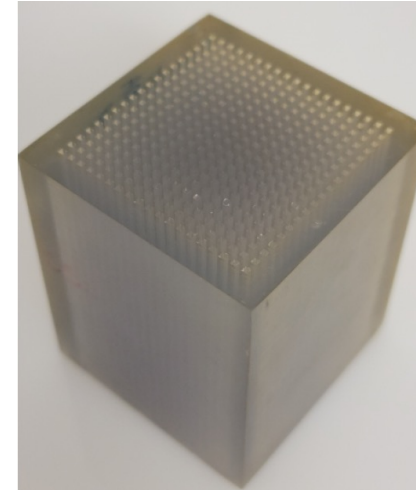
## HIMIT Checkout (Jones et al. [12])

- Two samples: CSQ3 and GE01
- Frequencies up to **6 kHz** (CSQ3)
- SPL up to **155 dB** (GE01)

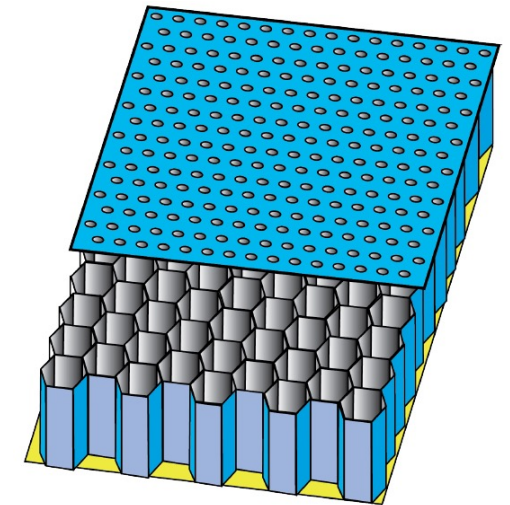
## Results

- Agreement between impedance of modes with highest power and ZKTL
- Increase in  $\theta$  of GE01 with SPL

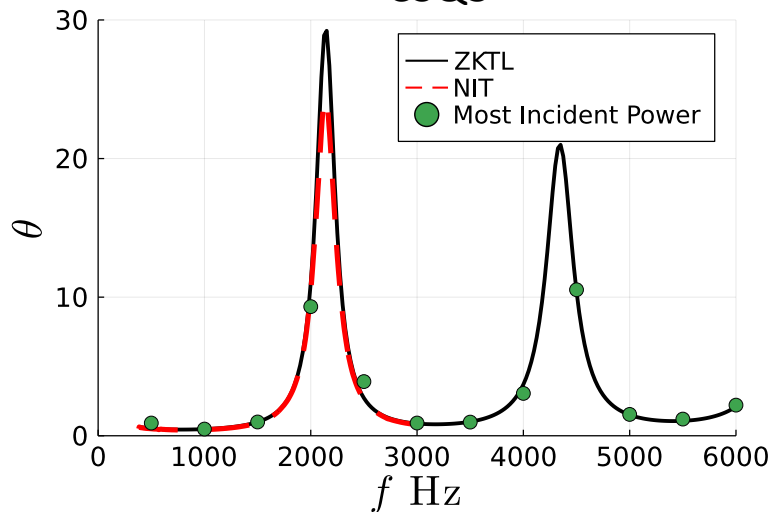
CSQ3



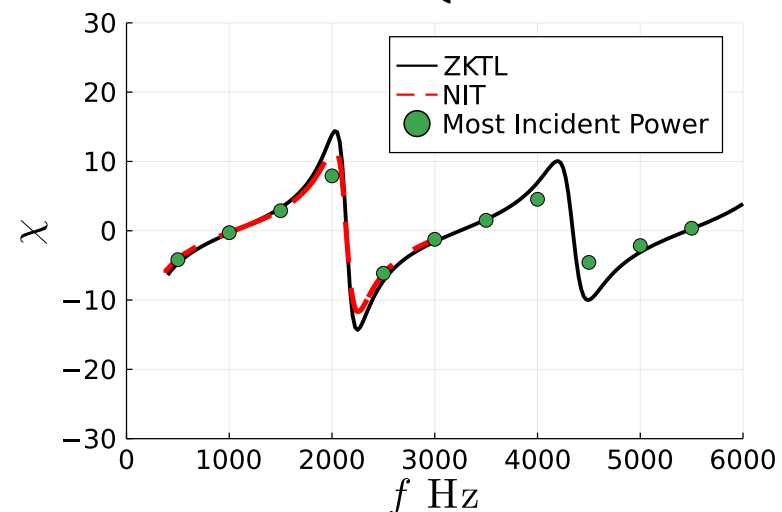
GE01



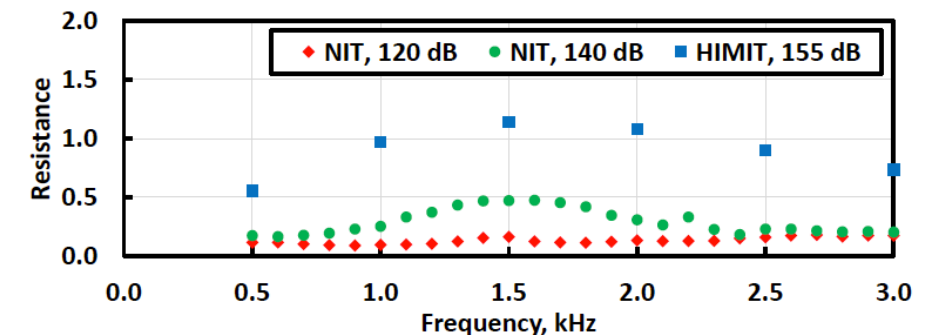
CSQ3



CSQ3



GE01





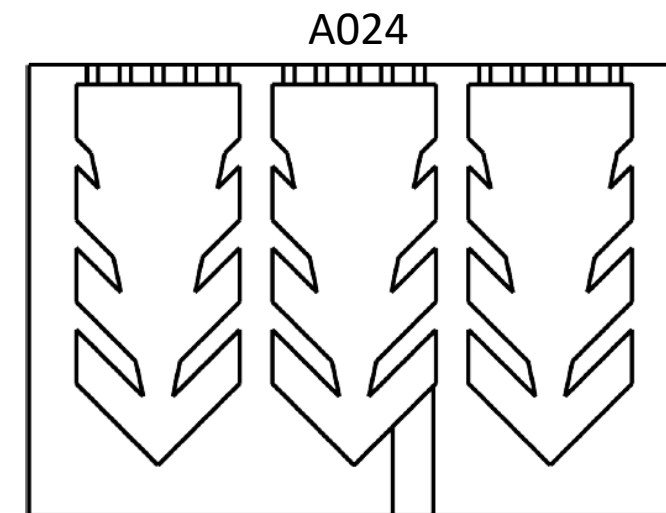
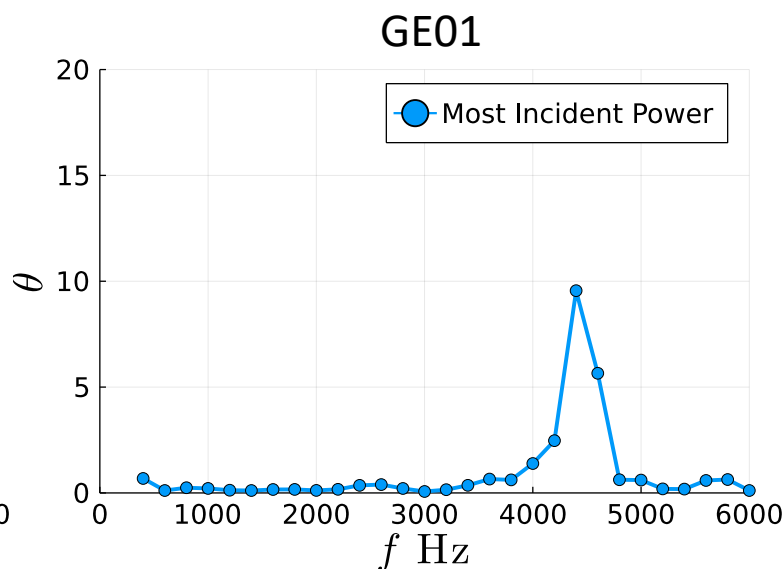
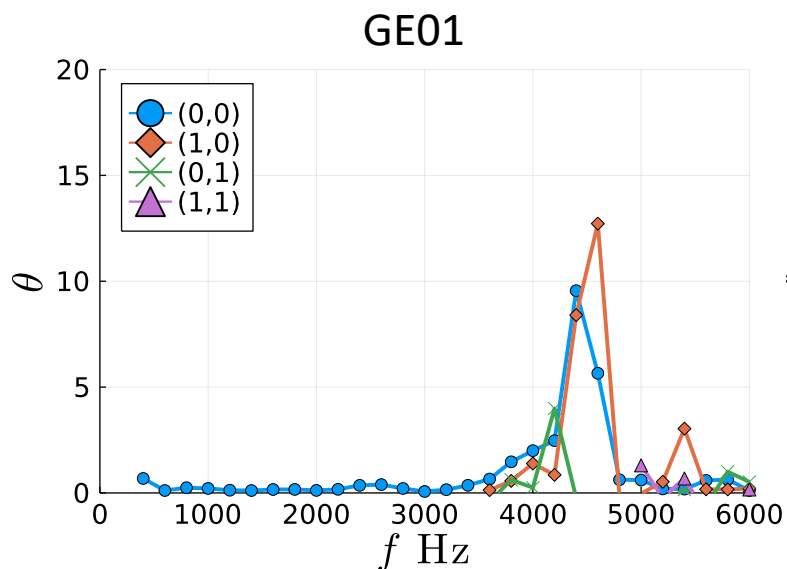
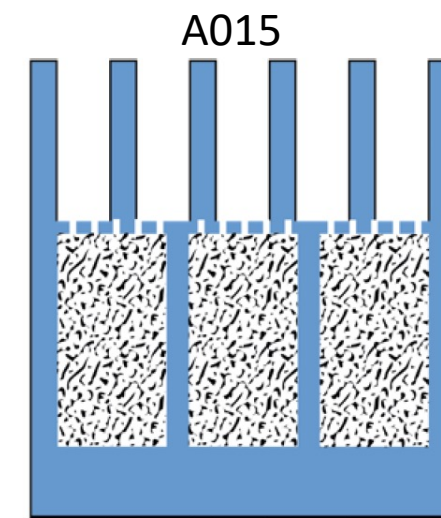
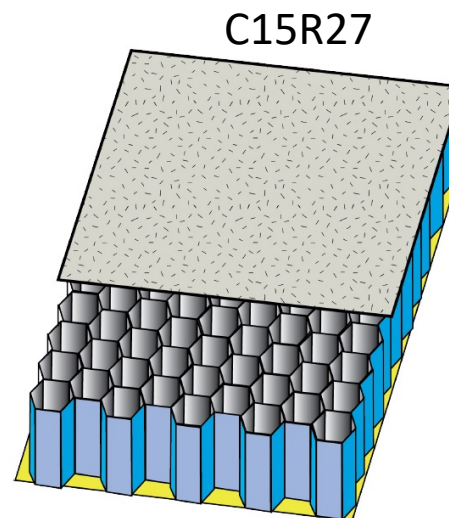
# Investigations with HIMIT (so far)

## Follow on Study (Solano et al. [13])

- C15R27: mesh over honeycomb
- A015 and A024: over the rotor samples
- 120 dB, 140 dB, and 150 dB
- $0.4 \leq f \leq 6.0$  kHz (150 dB;  $f \leq 3.0$  kHz)

## Results

- GE01 and A015
  - Highest power mode provides consistent results

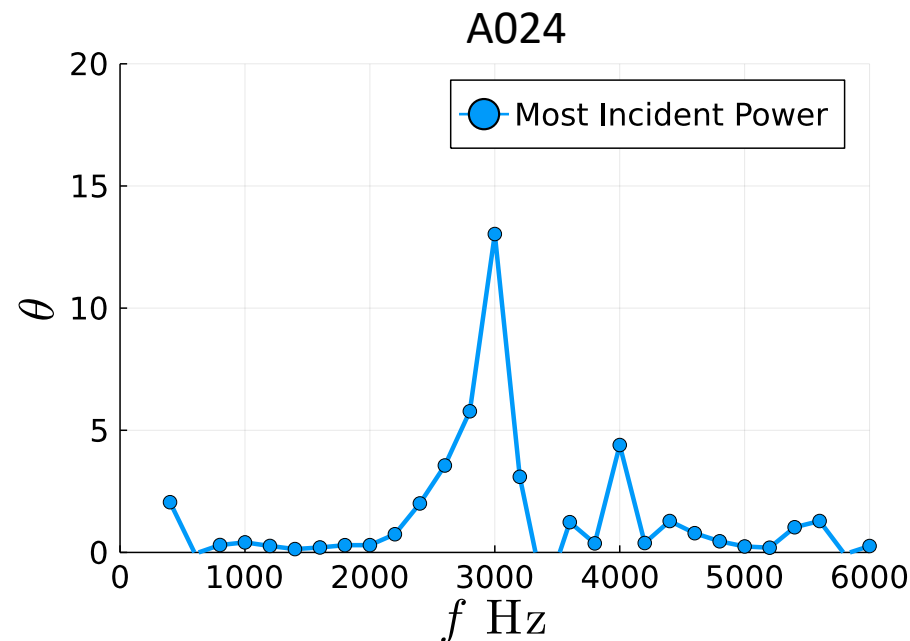
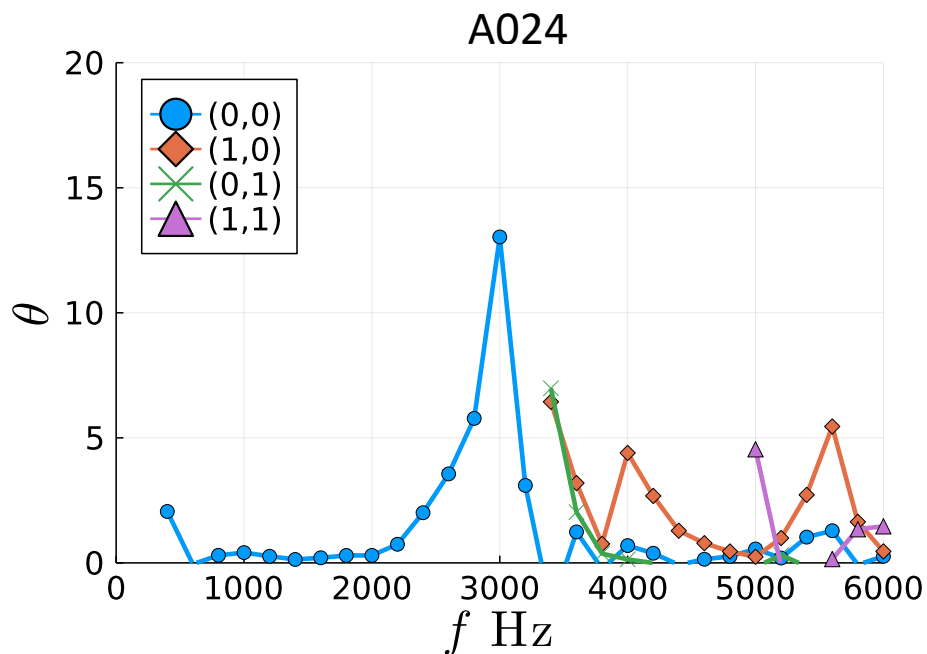
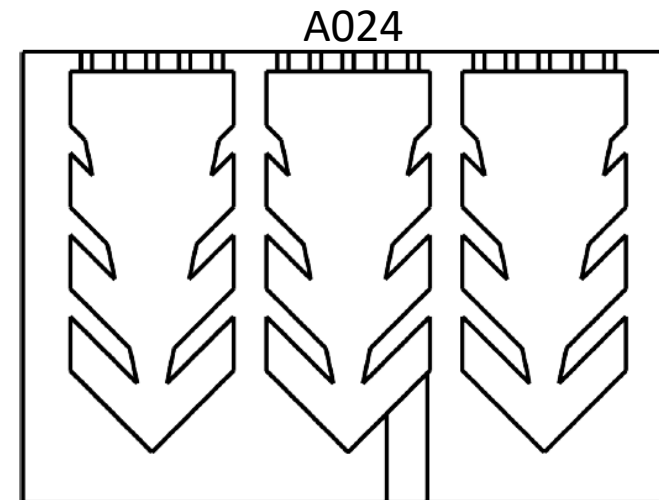




# Current Issues: A024 Sample

## Follow on Study (Solano et al. [13])

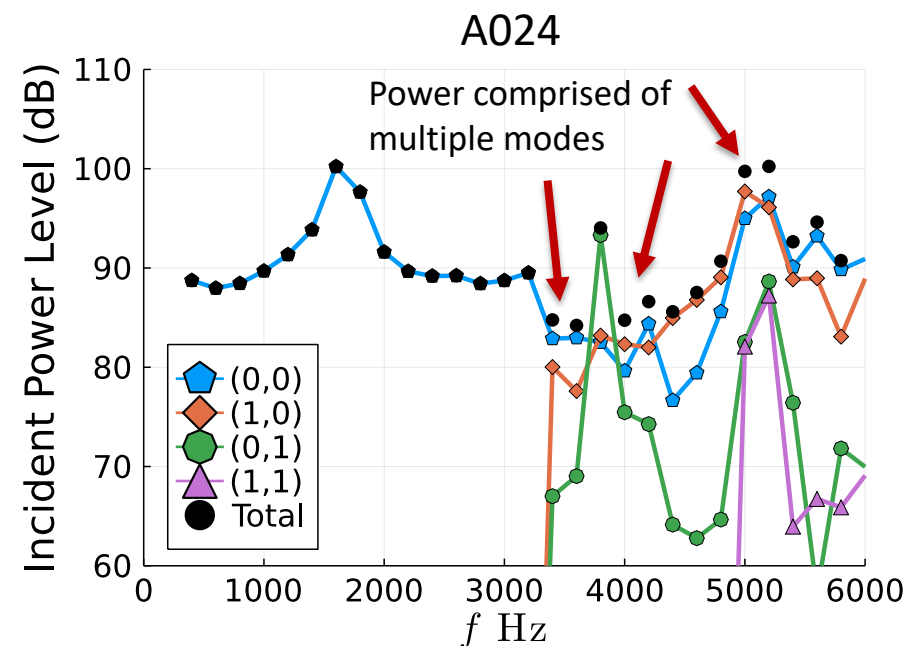
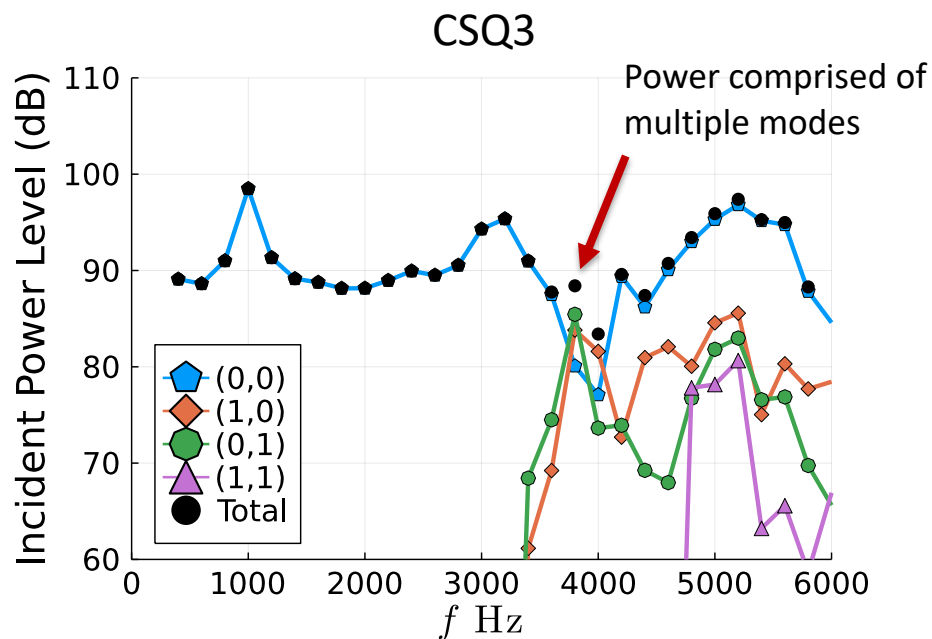
- A024 Results
  - Inconsistencies when using the highest power mode
  - Negative resistance values – nonphysical
  - Choosing mode with highest SPL does not help
  - COMSOL modeling?



# Current Issues: Mode Control

- Multiple modes contribute significantly around 4 kHz
- Modal decomposition approach works best when 1 mode is dominant
- Desire to have mode control (similar to CDTR) to avoid this issue
- Schultz et al. [10] – able to achieve 20 dB separation with blocker plate

(1,1) blocker plate



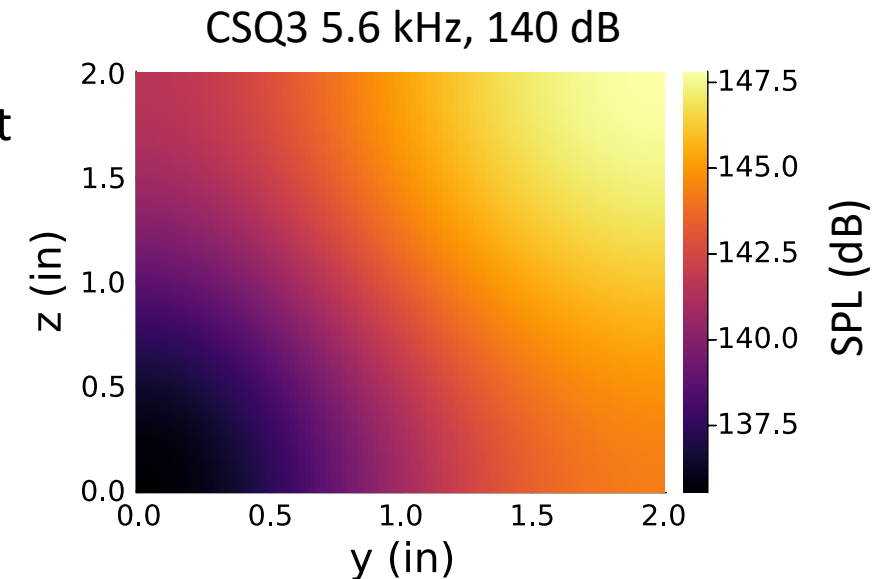
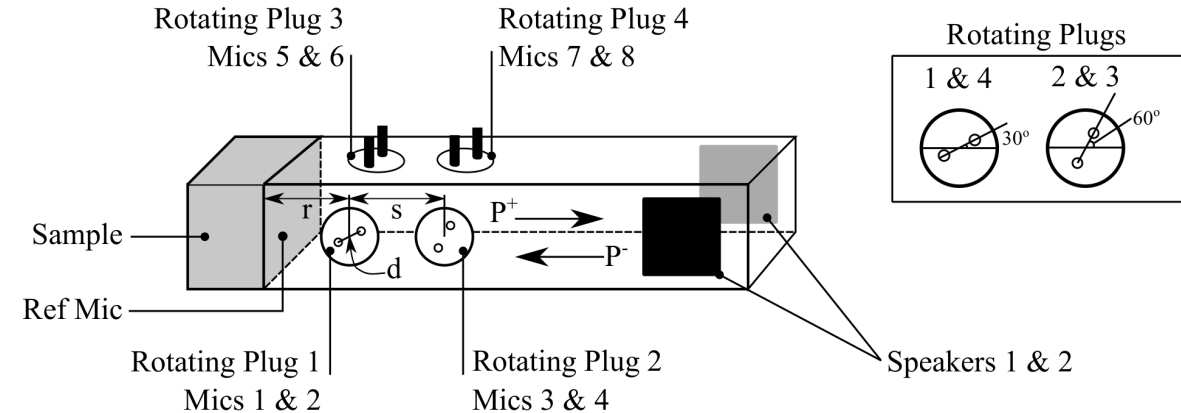
# Current Issues: Level Setting in HIMIT

## Reference microphone location

- Not ideal for HOMs
- $\frac{1}{4}$  inch microphone, as opposed to  $\frac{1}{8}$  inch plug mics
- Measuring SPL  $\frac{1}{4}$  inch away from sample

## Modal Decomposition Method to set level

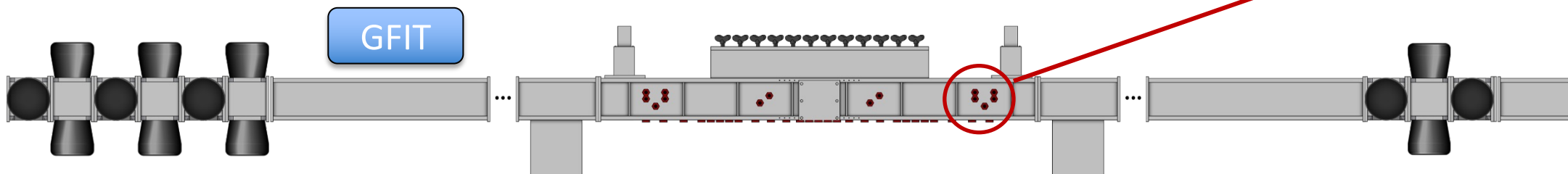
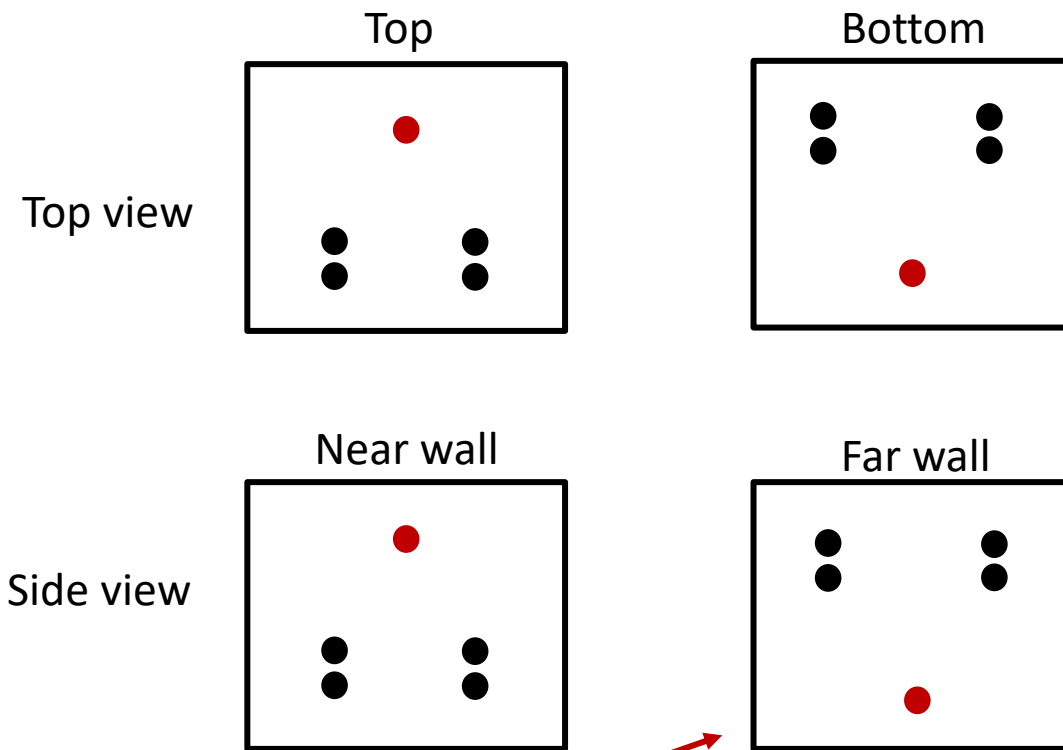
- May alleviate some of these difficulties
- 8 microphones to set the level
  - Use modal amplitudes to approximate SPL at face of sample
  - Insensitive to null locations with proper microphone placement
- Level setting options
  - Total, incident, or reflected **mode** power
  - Total, incident, or reflected **mode** SPL
  - **Average** total, incident, or reflected SPL at the sample face
- Implications for nonlinear samples? (perforate over honeycomb)

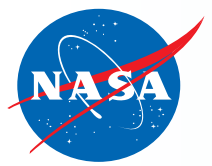


# Grazing Flow Impedance Tube (GFIT)

## High Frequency Testing in GFIT

- We intend to go up to **6 kHz** in GFIT by the end of the year
- Modal decomposition upstream and downstream
  - Need to utilize all mics in those arrays
- Likely need mode control in GFIT as well
- Impedance eduction [14]
  - Objective function method – CHE simulations in 3D
  - Prony – can handle vertical modes
  - Any desire for 3D shear flow with horizontal modes?
- In the meantime, examine mode attenuation





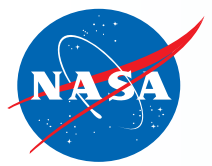
# Summary and Future Work

## Summary

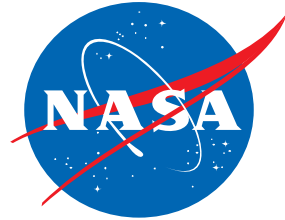
- Multiple samples have been tested in HIMIT up to 6 kHz and up to 155 dB
- Impedance deduced from mode with most incident power
- Our current analysis techniques work best when a single mode is dominant
  - Engine environments have many modes present
- Current obstacles
  - Setting level at sample face – SPL varies along cross-section

## Future Work

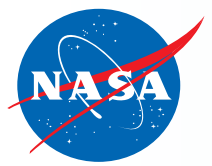
- Set level with modal decomposition method
- Mode control
- Investigate deduced impedance of HOMs at high amplitudes
- Implement modal decomposition in GFIT upstream and downstream arrays
  - Intend to go up to 6 kHz by the end of the year



# Questions?

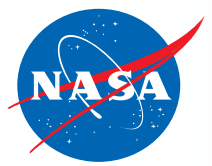






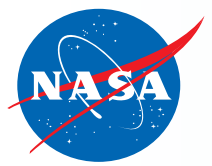
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# Backup Slides



# Background: NASA LaRC Liner Technology Facility

## High Intensity Modal Impedance Tube (HIMIT)

- Mach 0.0,  $\text{SPL} \leq 170$  dB,  $\text{Freq} \leq 6$  kHz
- Tone and Broadband sources

## Normal Incidence Tube (NIT)

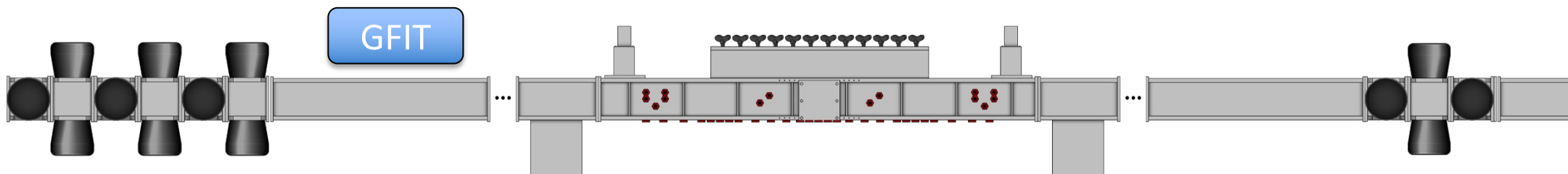
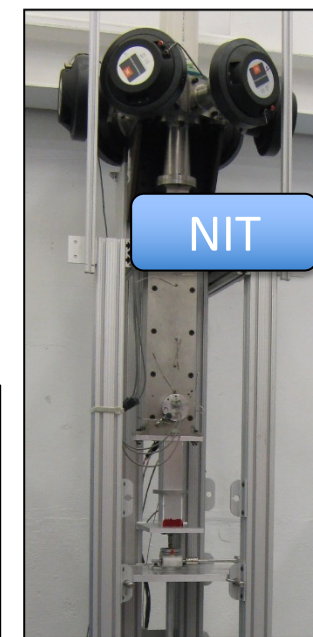
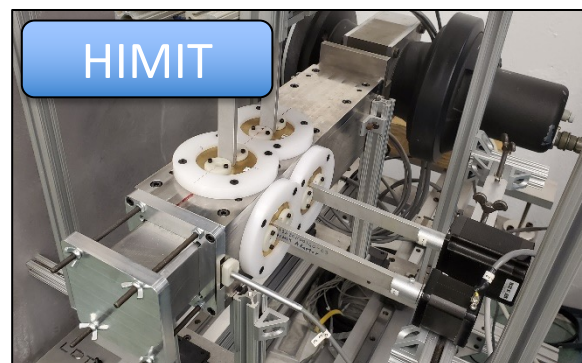
- Mach 0.0,  $\text{SPL} \leq 155$  dB,  $\text{Freq} \leq 3$  kHz
- Tone and Broadband sources

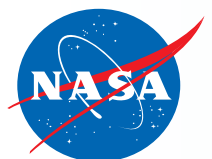
## Grazing Flow Impedance Tube (GFIT)

- Mach  $\leq 0.6$ ,  $\text{SPL} \leq 155$  dB,  $\text{Freq} \leq 3$  kHz
- Tone source

## Curved Duct Test Rig (CDTR)

- Mach  $\leq 0.5$ ,  $\text{SPL} \leq 135$  dB,  $\text{Freq} \leq 3$  kHz
- *Controlled* Tonal mode and Broadband sources





# Current Issues: Impedance of Higher Order Modes

## Perforate over Honeycomb

- Resistance proportional to rms acoustic velocity at high SPL
- Two parameter, Motsinger and Kraft [15], Crandall [16]

$$\theta_i = A + BV_i$$

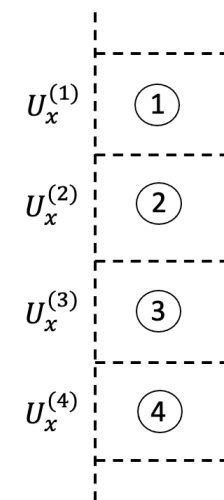
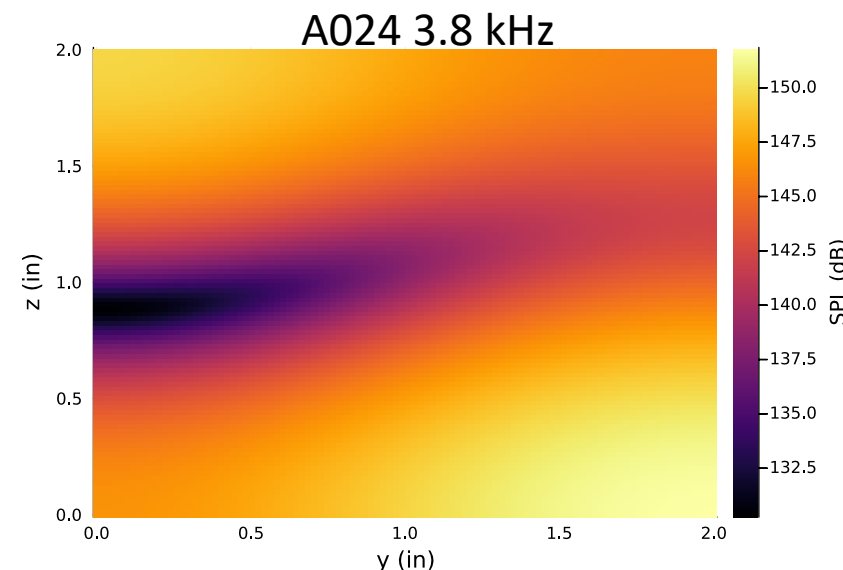
$$u_{rms}^{(i)} = \frac{|U_x^{(i)}|}{\sqrt{2}}$$

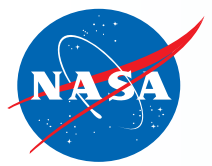
- Sample face will be exposed to varying SPL when HOMs dominate
- Resistance may need to be computed as an average over all the holes

$$\theta_{eff} = \frac{1}{N_h} \sum_{i=1}^{N_h} \theta_i = A + \frac{B}{N_h} \sum_{i=1}^{N_h} \frac{|U_x^{(i)}|}{\sqrt{2}}$$

- Replace the summation with an integral

$$\frac{1}{N_h} \sum_{i=1}^{N_h} \frac{|U_x^{(i)}|}{\sqrt{2}} \approx \frac{1}{ab} \int_0^b \int_0^a \frac{|U_x^{(i)}|}{\sqrt{2}} dy dz$$





# Preliminary Modification to Liner Models

- Eqn. 1 obtained with linear momentum relation

$$i\rho ckU_x = -\frac{\partial P}{\partial x}$$

- This breaks down near surface of liner
- Resort to liner models to understand what may happen when a HOM is dominant
- Two parameter, Mot and Kraft, Crandall

$$\theta_i = A + BV_i$$

- $V_i$  may be replaced by acoustic rms velocity

$$u_{rms}^{(i)} = \frac{|U_x^{(i)}|}{\sqrt{2}}$$

- The “effective” resistance of the liner may be the average of all the holes

$$\theta_{eff} = \frac{1}{N_h} \sum_{i=1}^{N_h} \theta_i = A + \frac{B}{N_h} \sum_{i=1}^{N_h} \frac{|U_x^{(i)}|}{\sqrt{2}}$$

- Replace the summation with an integral

$$\frac{1}{N_h} \sum_{i=1}^{N_h} \frac{|U_x^{(i)}|}{\sqrt{2}} \approx \frac{1}{ab} \int_0^b \int_0^a \frac{|U_x^{(i)}|}{\sqrt{2}} dy dz$$





# Preliminary Modification to Liner Models

- Consider the case where only one mode is dominant
- If this is the plane-wave mode, then

$$\theta_{eff}^{00} = A + \frac{B}{\rho c} \frac{|R_{00} - 1|}{\sqrt{2}} |A_{00}^-|$$

- The resistance for any HOM is

$$\theta_{eff}^{nm} = A + \frac{B}{\rho c} \frac{|\gamma_{nm}|}{k} \left( \frac{\sqrt{2}}{\pi} \right)^{\alpha_{nm}} |R_{nm} - 1| |A_{nm}^-|$$

- Where the following relations were used

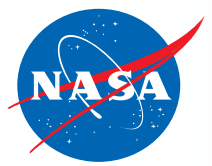
$$\frac{1}{ab} \int_0^b \int_0^a \frac{|\Psi_{01}|}{\sqrt{2}} dy dz = \frac{1}{ab} \int_0^b \int_0^a \frac{|\Psi_{10}|}{\sqrt{2}} dy dz = \frac{\sqrt{2}}{\pi}$$

$$\frac{1}{ab} \int_0^b \int_0^a \frac{|\Psi_{11}|}{\sqrt{2}} dy dz = \left( \frac{\sqrt{2}}{\pi} \right)^2$$

- Ratio of resistance of HOM to plane-wave mode

$$\frac{\theta_{eff}^{nm} - A}{\theta_{eff}^{00} - A} = \frac{|\gamma_{nm}|}{k} \left( \frac{2}{\pi} \right)^{\alpha_{nm}} \frac{|R_{nm} - 1|}{|R_{00} - 1|} \frac{|A_{nm}^-|}{|A_{00}^-|} \quad (2)$$

- It's possible that “effective” resistance is being measured when Eqn. 1 is used



# Preliminary Modification to Liner Models

- Bounds

$$\frac{|\gamma_{nm}|}{k} \leq 1 \quad \left(\frac{2}{\pi}\right)^{\alpha_{nm}} \leq 1$$

- Implies that

$$\frac{\theta_{eff}^{nm} - A}{\theta_{eff}^{00} - A} \leq \frac{|R_{nm} - 1| |A_{nm}^-|}{|R_{00} - 1| |A_{00}^-|}$$

- For the same incident level, will reflection coefficient adjust to keep resistance constant across different modes?